



## پردازش سیگنال دیجیتال

درس ۲۳

# تبديل فوريه‌ي گسته

The Discrete Fourier Transform

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<http://courses.fouladi.ir/dsp>

## The Discrete Fourier Transform

# Sampling the Fourier Transform

- Consider an **aperiodic** sequence with a Fourier transform

$$x[n] \xleftarrow{\text{DTFT}} X(e^{j\omega})$$

- Assume that a sequence is obtained by **sampling** the DTFT

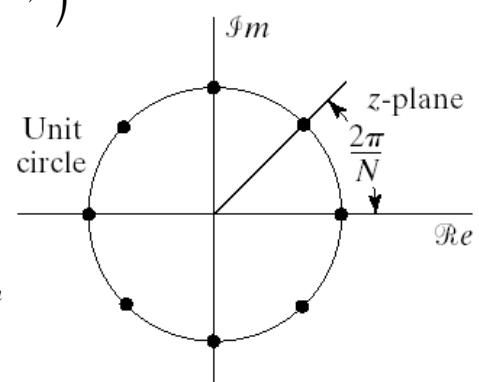
$$\tilde{X}[k] = X(e^{j\omega}) \Big|_{\omega=(2\pi/N)k} = X(e^{j(2\pi/N)k})$$

- Since the DTFT is periodic resulting sequence is also periodic
- We can also write it in terms of the z-transform

$$\tilde{X}[k] = X(z) \Big|_{z=e^{j(2\pi/N)k}} = X(e^{j(2\pi/N)k})$$

- The sampling points are shown in figure
- $\tilde{X}[k]$  could be the DFS of a sequence
- Write the corresponding sequence

$$\tilde{x}[n] = \frac{1}{N} \sum_{k=0}^{N-1} \tilde{X}[k] e^{j(2\pi/N)kn}$$



## Sampling the Fourier Transform Cont'd

- The only assumption made on the sequence is that DTFT exists

$$X(e^{j\omega}) = \sum_{m=-\infty}^{\infty} x[m]e^{-j\omega m} \quad \tilde{X}[k] = X\left(e^{j(2\pi/N)k}\right) \quad \tilde{x}[n] = \frac{1}{N} \sum_{k=0}^{N-1} \tilde{X}[k] e^{j(2\pi/N)kn}$$

- Combine equation to get

$$\begin{aligned} \tilde{x}[n] &= \frac{1}{N} \sum_{k=0}^{N-1} \left[ \sum_{m=-\infty}^{\infty} x[m] e^{-j(2\pi/N)km} \right] e^{j(2\pi/N)kn} \\ &= \sum_{m=-\infty}^{\infty} x[m] \left[ \frac{1}{N} \sum_{k=0}^{N-1} e^{j(2\pi/N)k(n-m)} \right] = \sum_{m=-\infty}^{\infty} x[m] \tilde{p}[n-m] \end{aligned}$$

- Term in the parenthesis is

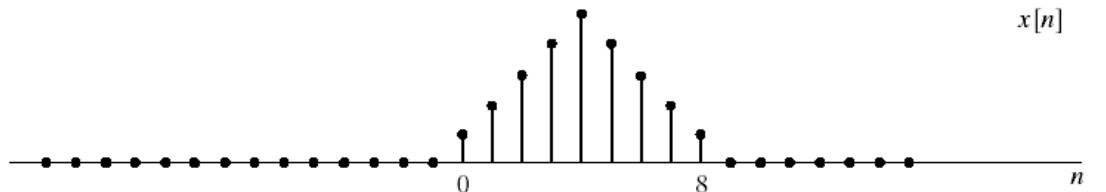
$$\frac{1}{N} \sum_{n=0}^{N-1} e^{j(2\pi/N)(k-r)n} = \begin{cases} 1, & k-r = mN, \quad m \text{ an integer}, \\ 0, & \text{otherwise}. \end{cases}$$

$$\tilde{p}[n-m] = \frac{1}{N} \sum_{k=0}^{N-1} e^{j(2\pi/N)k(n-m)} = \sum_{r=-\infty}^{\infty} \delta[n-m-rN]$$

- So we get

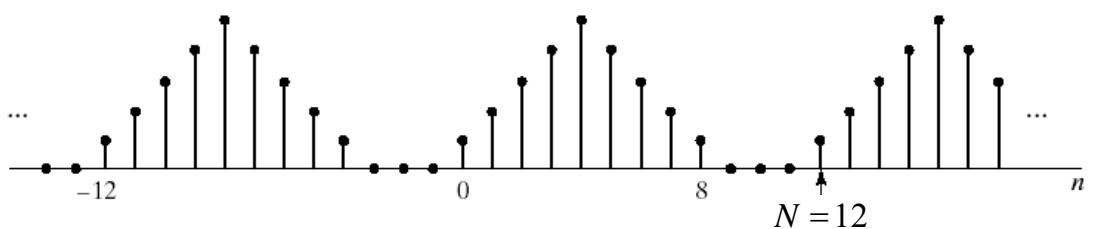
$$\tilde{x}[n] = x[n] * \sum_{r=-\infty}^{\infty} \delta[n-rN] = \sum_{r=-\infty}^{\infty} x[n-rN]$$

## Sampling the Fourier Transform Cont'd



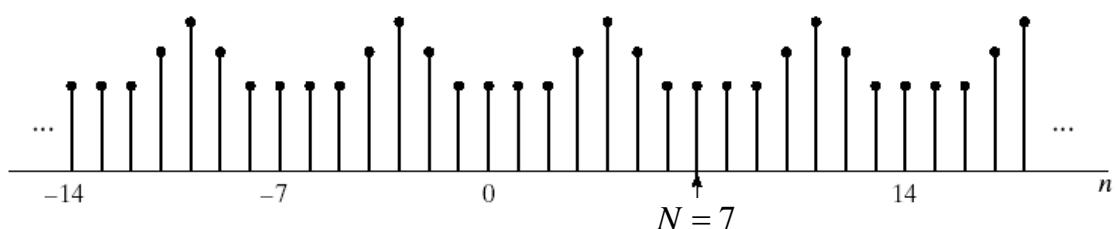
(a)

$$\tilde{x}[n] = \sum_{r=-\infty}^{\infty} x[n - r]12$$



$N=12$

$$\tilde{x}[n] = \sum_{r=-\infty}^{\infty} x[n - r]7$$



$N=7$

## Sampling the Fourier Transform Cont'd

- Samples of the DTFT of an **aperiodic sequence**
  - can be thought of as **DFS coefficients** of a **periodic sequence**
  - obtained through summing periodic replicas of original sequence
- If the original sequence
  - is of **finite length**
  - and we take **sufficient number of samples of its DTFT**
  - the original sequence can be recovered by

$$x[n] = \begin{cases} \tilde{x}[n] & 0 \leq n \leq N-1 \\ 0 & \text{otherwise} \end{cases}$$

- It is not necessary to know the DTFT at all frequencies
  - to recover the discrete-time sequence in time domain
- **Discrete Fourier Transform**
  - Representing a finite length sequence by **samples of DTFT**

# The Discrete Fourier Transform

- Consider a **finite length** sequence  $x[n]$  of length  $N$

$$x[n] = 0 \text{ outside of } 0 \leq n \leq N-1$$

- For given length- $N$  sequence associate a periodic sequence

$$\tilde{x}[n] = \sum_{r=-\infty}^{\infty} x[n-rN]$$

- The **DFS coefficients** of the **periodic sequence** are **samples** of the **DTFT** of  $x[n]$
- Since  $x[n]$  is of length  $N$  there is no overlap between terms of  $x[n-rN]$  and we can write the periodic sequence as

$$\tilde{x}[n] = x[(n \bmod N)] = x[((n))_N]$$

- To maintain **duality** between **time** and **frequency**

- We choose **one period** of  $\tilde{X}[k]$  as the Fourier transform of  $x[n]$

$$X[k] = \begin{cases} \tilde{X}[k] & 0 \leq k \leq N-1 \\ 0 & \text{otherwise} \end{cases}$$

$$\tilde{X}[k] = X[(k \bmod N)] = X[((k))_N]$$

## The Discrete Fourier Transform Cont'd

- The DFS pair

$$\tilde{X}[k] = \sum_{n=0}^{N-1} \tilde{x}[n] e^{-j(2\pi/N)kn}$$
$$\tilde{x}[n] = \frac{1}{N} \sum_{k=0}^{N-1} \tilde{X}[k] e^{j(2\pi/N)kn}$$

- The summations involve only **one period** so we can write:

$$X[k] = \begin{cases} \sum_{n=0}^{N-1} \tilde{x}[n] e^{-j(2\pi/N)kn} & 0 \leq k \leq N-1 \\ 0 & \text{otherwise} \end{cases}$$
$$x[n] = \begin{cases} \frac{1}{N} \sum_{k=0}^{N-1} \tilde{X}[k] e^{j(2\pi/N)kn} & 0 \leq n \leq N-1 \\ 0 & \text{otherwise} \end{cases}$$

- The Discrete Fourier Transform

$$X[k] = \sum_{n=0}^{N-1} x[n] e^{-j(2\pi/N)kn}$$
$$x[n] = \frac{1}{N} \sum_{k=0}^{N-1} X[k] e^{j(2\pi/N)kn}$$

- The DFT pair can also be written as

$$x[n] \xleftarrow{\text{DFT}} X[k]$$

## The Discrete Fourier Transform Cont'd

- The Discrete Fourier Transform

$$X[k] = \sum_{n=0}^{N-1} x[n] e^{-j(2\pi/N)kn}$$

$$x[n] = \frac{1}{N} \sum_{k=0}^{N-1} X[k] e^{j(2\pi/N)kn}$$

$$X[k] = \sum_{n=0}^{N-1} x[n] W_N^{kn}$$

$$x[n] = \frac{1}{N} \sum_{k=0}^{N-1} X[k] W_N^{-kn}$$

$$k = 1, 2, \dots, N - 1$$

$$n = 1, 2, \dots, N - 1$$

- The DFT pair can also be written as

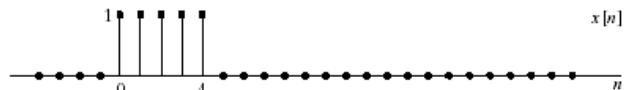
$$x[n] \xrightarrow{\text{DFT}} X[k]$$

# Example

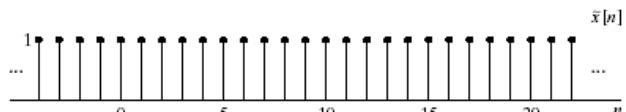
The DFT of a rectangular pulse

- $x[n]$  is of length 5
- We can consider  $x[n]$  of any length greater than or equal to 5
- Let's pick  $N = 5$
- Calculate the DFS of the periodic form of  $x[n]$

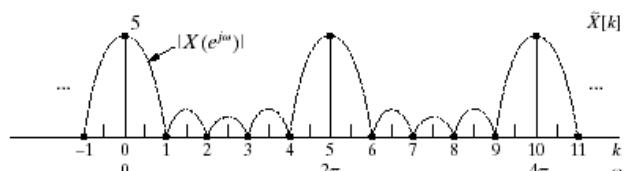
$$\begin{aligned}\widetilde{X}[k] &= \sum_{n=0}^4 e^{-j(2\pi k/5)n} \\ &= \frac{1 - e^{-j2\pi k}}{1 - e^{-j(2\pi k/5)}} \\ &= \begin{cases} 5 & k = 0, \pm 5, \pm 10, \dots \\ 0 & \text{otherwise} \end{cases}\end{aligned}$$



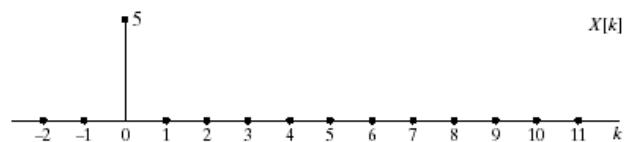
(a)



(b)



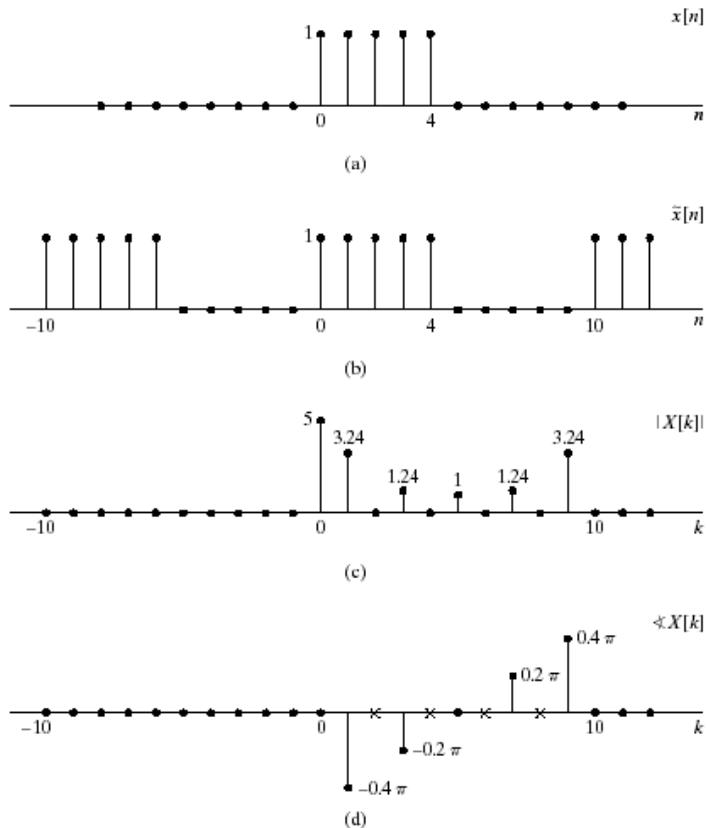
(c)



## Example Cont'd

The DFT of a rectangular pulse

- If we consider  $x[n]$  of length 10
  - We get a different set of DFT coefficients
- Still samples of the DTFT but in different places



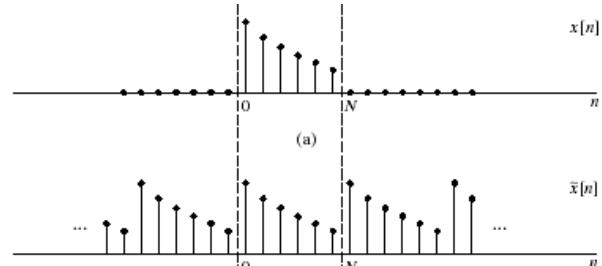
# Properties of DFT

- **Linearity**

$$x_1[n] \quad \xleftarrow{DFT} \quad X_1[k]$$

$$x_2[n] \quad \xleftarrow{DFT} \quad X_2[k]$$

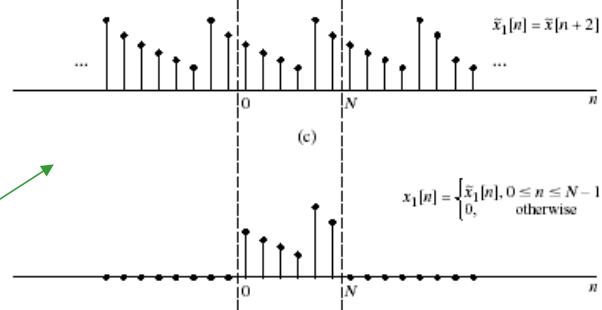
$$ax_1[n] + bx_2[n] \quad \xleftarrow{DFT} \quad aX_1[k] + bX_2[k]$$



- **Duality**

$$x[n] \quad \xleftarrow{DFT} \quad X[k]$$

$$X[n] \quad \xleftarrow{DFT} \quad Nx[((-k))_N]$$

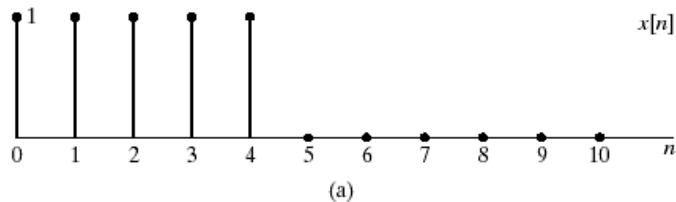


- **Circular Shift of a Sequence**

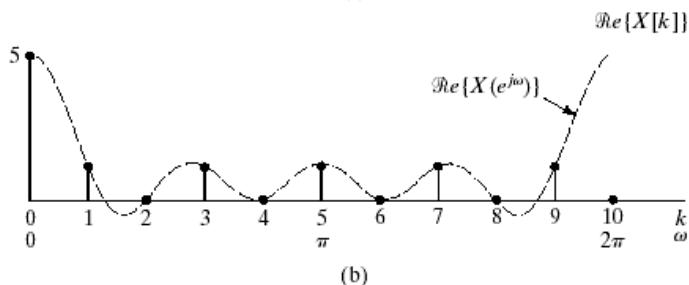
$$x[n] \quad \xleftarrow{DFT} \quad X[k]$$

$$x[((n-m))_N] \quad 0 \leq n \leq N-1 \quad \xleftarrow{DFT} \quad X[k]e^{-j(2\pi k/N)m}$$

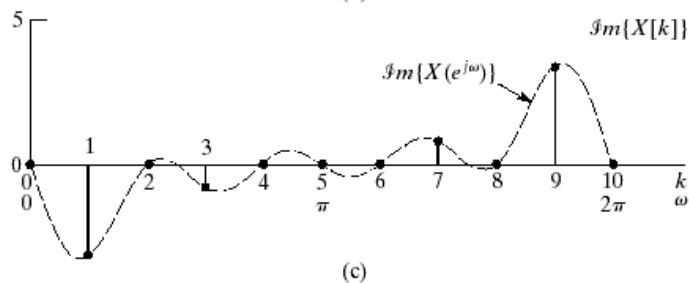
# Example: Duality



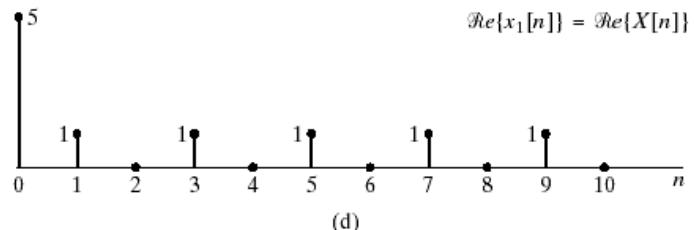
$x[n]$



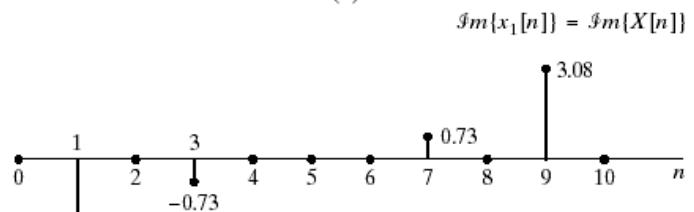
$\Re e\{X[k]\}$



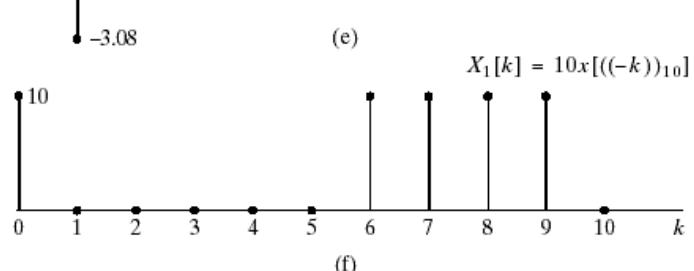
$\Im m\{X[k]\}$



$\Re e\{x_1[n]\} = \Re e\{X[n]\}$



$\Im m\{x_1[n]\} = \Im m\{X[n]\}$



$X_1[k] = 10x[(-k)]_{10}$

# Properties of DFT

Finite-Length Sequence (Length $N$ )	$N$ -point DFT (Length $N$ )
1. $x[n]$	$X[k]$
2. $x_1[n], x_2[n]$	$X_1[k], X_2[k]$
3. $ax_1[n] + bx_2[n]$	$aX_1[k] + bX_2[k]$
4. $X[n]$	$Nx[((-k))_N]$
5. $x[((n - m))_N]$	$W_N^{km} X[k]$
6. $W_N^{-\ell n} x[n]$	$X[((k - \ell))_N]$
7. $\sum_{m=0}^{N-1} x_1[m]x_2[((n - m))_N]$	$X_1[k]X_2[k]$
8. $x_1[n]x_2[n]$	$\frac{1}{N} \sum_{\ell=0}^{N-1} X_1[\ell]X_2[((k - \ell))_N]$
9. $x^*[n]$	$X^*((-k))_N]$
10. $x^*[((-n))_N]$	$X^*[k]$

## Properties of DFT Cont'd

Finite-Length Sequence (Length  $N$ )

$$11. \quad \mathcal{R}e\{x[n]\}$$

$$12. \quad j\mathcal{I}m\{x[n]\}$$

$$13. \quad x_{\text{ep}}[n] = \frac{1}{2}\{x[n] + x^*[((-n))_N]\}$$

$$14. \quad x_{\text{op}}[n] = \frac{1}{2}\{x[n] - x^*[((-n))_N]\}$$

Properties 15–17 apply only when  $x[n]$  is real.

15. Symmetry properties

$$16. \quad x_{\text{ep}}[n] = \frac{1}{2}\{x[n] + x[((-n))_N]\}$$

$$17. \quad x_{\text{op}}[n] = \frac{1}{2}\{x[n] - x[((-n))_N]\}$$

$N$ -point DFT (Length  $N$ )

$$X_{\text{ep}}[k] = \frac{1}{2}\{X[((k))_N] + X^*[((-k))_N]\}$$

$$X_{\text{op}}[k] = \frac{1}{2}\{X[((k))_N] - X^*[((-k))_N]\}$$

$$\mathcal{R}e\{X[k]\}$$

$$j\mathcal{I}m\{X[k]\}$$

$$\begin{cases} X[k] = X^*[((-k))_N] \\ \mathcal{R}e\{X[k]\} = \mathcal{R}e\{X[((-k))_N]\} \\ \mathcal{I}m\{X[k]\} = -\mathcal{I}m\{X[((-k))_N]\} \\ |X[k]| = |X[((-k))_N]| \\ \angle\{X[k]\} = -\angle\{X[((-k))_N]\} \end{cases}$$

$$\mathcal{R}e\{X[k]\}$$

$$j\mathcal{I}m\{X[k]\}$$

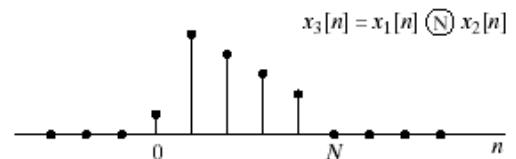
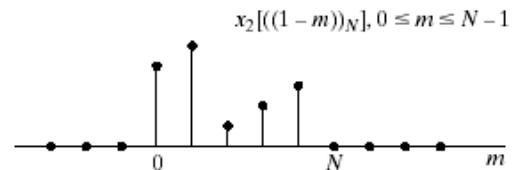
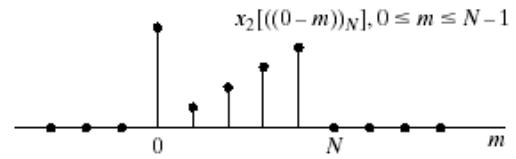
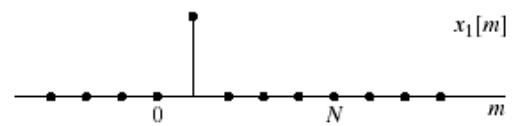
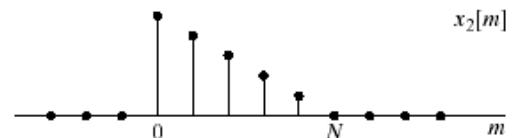
# Circular Convolution

- Circular convolution of two finite length sequences

$$x_3[n] = x_1[n] \circledast x_2[n]$$

$$x_3[n] = \sum_{m=0}^{N-1} x_1[m] x_2[((n-m))_N]$$

$$x_3[n] = \sum_{m=0}^{N-1} x_2[m] x_1[((n-m))_N]$$



## Example

- Circular convolution of two rectangular pulses  $L = N = 6$

$$x_1[n] = x_2[n] = \begin{cases} 1 & 0 \leq n \leq L-1 \\ 0 & \text{otherwise} \end{cases}$$

- DFT of each sequence

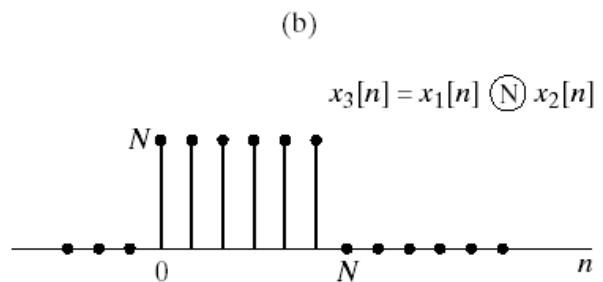
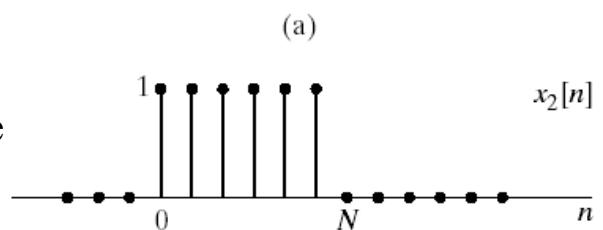
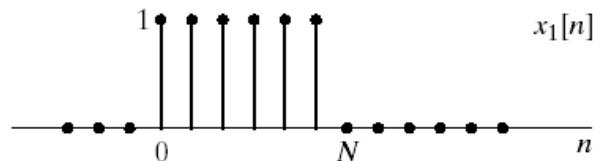
$$X_1[k] = X_2[k] = \sum_{n=0}^{N-1} e^{-j\frac{2\pi}{N}kn} = \begin{cases} N & k = 0 \\ 0 & \text{otherwise} \end{cases}$$

- Multiplication of DFTs

$$X_3[k] = X_1[k]X_2[k] = \begin{cases} N^2 & k = 0 \\ 0 & \text{otherwise} \end{cases}$$

- And the inverse DFT

$$x_3[n] = \begin{cases} N & 0 \leq n \leq N-1 \\ 0 & \text{otherwise} \end{cases}$$



## Example

- We can augment zeros to each sequence  $N = 2L = 12$
- The DFT of each sequence

$$X_1[k] = X_2[k] = \frac{1 - e^{-j\frac{2\pi Lk}{N}}}{1 - e^{-j\frac{2\pi k}{N}}}$$

- Multiplication of DFTs

$$X_3[k] = \left( \frac{1 - e^{-j\frac{2\pi Lk}{N}}}{1 - e^{-j\frac{2\pi k}{N}}} \right)^2$$

