



بردازش سیگنال دیجیتال

درس ۴

تبديل فوريه گسسته-زمان

## **Discrete-Time Fourier Transform**

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http://courses.fouladi.ir/dsp

# **Frequency Response**

- The **frequency response** defines a systems output
  - for complex exponential at all frequencies
- If input signals can be represented as a sum of complex exponentials

$$x[n] = \sum_{k} \alpha_{k} e^{j \omega_{k} n}$$

- we can determine the output of the system

$$y[n] = \sum_{k} \alpha_{k} H(e^{j\omega_{k}}) e^{j\omega_{k}n}$$

- Different from continuous-time frequency response
  - Discrete-time frequency response is periodic with  $2\pi$

$$H(e^{j(\omega+2\pi r)}) = \sum_{k=-\infty}^{\infty} h[k] e^{-j(\omega+2\pi r)k} = \sum_{k=-\infty}^{\infty} h[k] e^{-j2\pi rk} e^{-j\omega k} = \sum_{k=-\infty}^{\infty} h[k] e^{-j\omega k}$$
$$H(e^{j(\omega+2\pi r)}) = H(e^{j\omega})$$

# **Discrete-Time Fourier Transform**

Many sequences can be expressed as a weighted sum of complex exponentials as

$$x[n] = \frac{1}{2\pi} \int_{-\pi}^{\pi} X(e^{j\omega}) e^{j\omega n} d\omega \quad \text{(inverse transform)}$$

• Where the weighting is determined as

$$X(e^{j\omega}) = \sum_{n=-\infty}^{\infty} x[n]e^{-j\omega n}$$
 (forward transform)

- $X(e^{j\omega})$  is the Fourier spectrum of the sequence x[n]
  - It specifies the magnitude and phase of the sequence
  - The phase wraps at  $2\pi$  hence is not uniquely specified
- The frequency response of a LTI system is the DTFT of the impulse response

$$H(e^{j\omega}) = \sum_{k=-\infty}^{\infty} h[k] e^{-j\omega k} \text{ and } h[n] = \frac{1}{2\pi} \int_{-\pi}^{\pi} H(e^{j\omega}) e^{j\omega n} d\omega$$

#### **Discrete-Time Fourier Transform Pair**

$$X(e^{j\omega}) = \sum_{n=-\infty}^{\infty} x[n] e^{-j\omega n} \text{ and } x[n] = \frac{1}{2\pi} \int_{-\pi}^{\pi} X(e^{j\omega}) e^{j\omega n} d\omega$$

- Let's show that they constitute a transform pair
  - Substitute first equation into second to get

$$\hat{x}[n] = \frac{1}{2\pi} \int_{-\pi}^{\pi} \left( \sum_{m=-\infty}^{\infty} x[m] e^{-j\omega m} \right) e^{j\omega n} d\omega = \sum_{m=-\infty}^{\infty} x[m] \left( \frac{1}{2\pi} \int_{-\pi}^{\pi} e^{j\omega(n-m)} d\omega \right)$$

- Evaluate the integral

$$\frac{1}{2\pi} \int_{-\pi}^{\pi} e^{j\omega(n-m)} d\omega = \begin{cases} \frac{\sin[\pi(n-m)]}{\pi(n-m)} = 0 & \text{for } n \neq m \\ 1 & \text{for } n = m \end{cases}$$
$$= \delta[n-m]$$

- Substitute the integral with this result to get

$$\hat{x}[n] = \sum_{m=-\infty}^{\infty} x[m] \delta[n-m] = x[n]$$

**Digital Signal Processing** 

$$X(e^{j\omega}) = \sum_{n=-\infty}^{\infty} x[n] e^{-j\omega n}$$

• For a given sequence the DTFT exist if the infinite sum convergence

$$|X(e^{j\omega})| < \infty$$
 for all  $\omega$ 

# • Or $\left| X(e^{j\omega}) \right| = \left| \sum_{n=-\infty}^{\infty} x[n] e^{-j\omega n} \right| \le \sum_{n=-\infty}^{\infty} |x[n]| e^{-j\omega n} = \sum_{n=-\infty}^{\infty} |x[n]| < \infty$

- So the DTFT exists if a given sequence is absolute summable
- All stable systems are absolute summable and have DTFTs

## **DTFT Demo**

Square Wave

Triangular Wave

From *Fundamentals of Signals and Systems Using the Web and Matlab* by Edward W. Kamen and Bonnie S. Heck